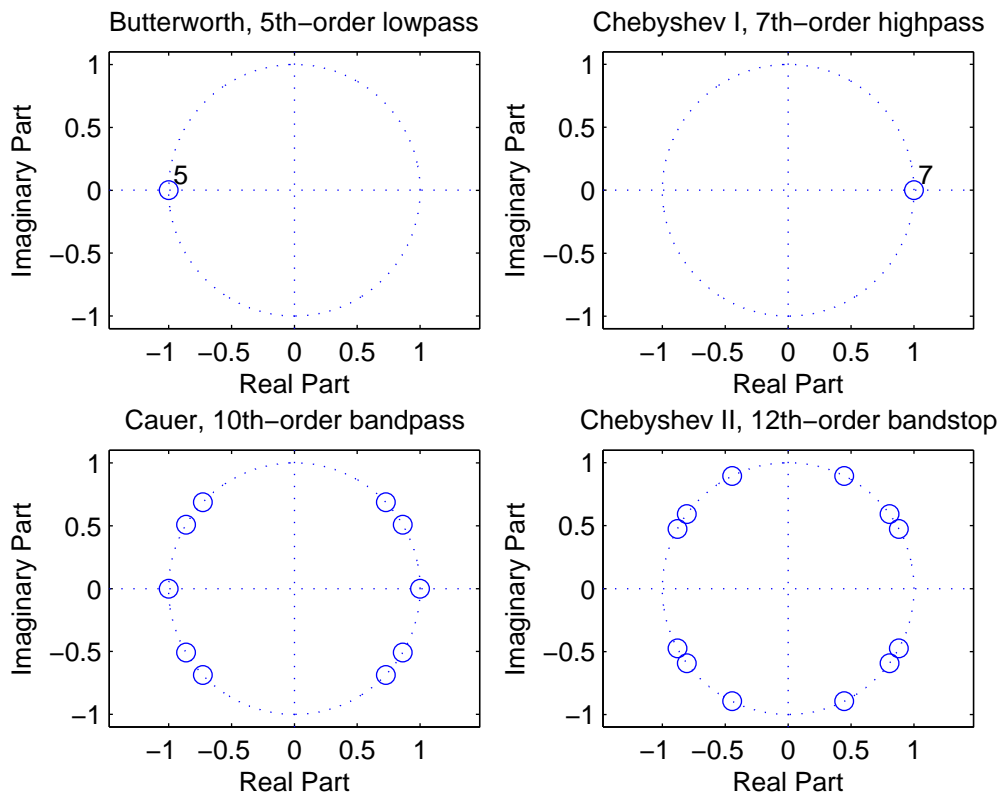


- 1
- a) Lowpass, because $H(1) = 4$, $H(-1)=0$
 - b) In two's-complementary arithmetics, temporary overflows are okay in partial sums if the final sum is within the number range \Rightarrow only inputs to noninteger multipliers and the output need to be scaled. Safe scaling: never overflow. Lp-norm scaling: overflow with a certain probability.
 - c) 1) Inserted unit elements between two-ports, propagated into the filter using Kuroda's identities. 2) Directly interconnected adaptors with reflection-free ports.
 - d) The group delay is $10T$ seconds (10 samples).
 - e)



- 2) Band edges: $\omega_c T_{\text{sample}} = 0.5 \pi$, $\omega_s T_{\text{sample}} = 0.84559 \pi$.

Bilinear transform $\Rightarrow \omega_{ac} = 1$, $\omega_{as} = 4.0417$.

Requirements on ω_{ac} , ω_{as} , and $A_{\text{min}} \Rightarrow 15 \leq \Theta \leq 15$.

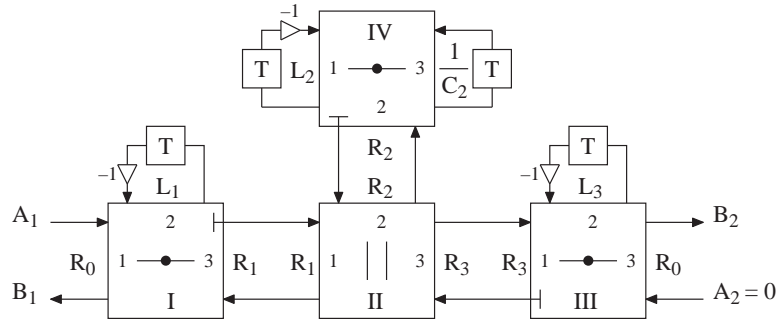
Normalized element values

$$L'_1 = L'_3 = 0.9944, L'_2 = 0.0463, C'_2 = 1.0941$$

Denormalize with $R_0 = 1$, $\omega_0 = \omega_{ac} = 1 \Rightarrow$

$$L_1 = L_3 = 0.9944, L_2 = 0.0463, C_2 = 1.0941$$

Wave flow graph



Adaptor coefficients:

$$\begin{aligned}
 R_1 &= L_1 + R_0 = 1.9944 & R_3 &= L_3 + R_0 = 1.9944 \\
 \text{I: } \alpha_1 &= \frac{2R_0}{R_0 + L_1 + R_1} = 0.5014 & \alpha_1 &= 1 \\
 \alpha_2 &= 1 - \alpha_1 = 0.4986 & \text{, III: } \alpha_2 &= \frac{2L_3}{R_3 + L_3 + R_0} = 0.4986 \\
 \alpha_3 &= 1 & \alpha_3 &= 1 - \alpha_2 = 0.5014 \\
 \\
 R_2 &= L_2 + \frac{1}{C_2} = 0.9603 & \alpha_1 &= \frac{\frac{2}{R_1}}{\frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3}} = 0.49057 \\
 \text{IV: } \alpha_1 &= \frac{2L_2}{L_2 + R_2 + \frac{1}{C_2}} = 0.039618 & \text{, II: } \alpha_2 &= \frac{\frac{2}{R_2}}{\frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3}} = 1.0189 \\
 \alpha_2 &= 1 & \alpha_3 &= 1 - \alpha_1 - \alpha_2 = 0.49057 \\
 \alpha_3 &= 1 - \alpha_1 = 0.60382 & &
 \end{aligned}$$

3) Requirements for analog HP filter:

$$\omega_{ac} = \frac{2}{T} \tan\left(\frac{\omega_c T}{2}\right) = \frac{2}{T} \tan(0,325\pi) \approx \frac{2}{T}(1, 631852)$$

$$\omega_{as} = \frac{2}{T} \tan\left(\frac{\omega_s T}{2}\right) = \frac{2}{T} \tan(0,2\pi) = \frac{2}{T}(0, 726542)$$

Requirements for analog LP prototype filter:

$$\Omega_c = \frac{\omega_I^2}{\omega_{ac}} \quad \Omega_s = \frac{\omega_I^2}{\omega_{as}}$$

Choose $\Omega_c = 1 \Rightarrow$

$$\omega_I^2 = \omega_{ac} \quad \Omega_s = \frac{\omega_{ac}}{\omega_{as}} \approx 2,24605$$

Order: nomogram $\Rightarrow N = 4$

Normalized poles:

$$S'_{1,2} = -0,1395360 \pm j0,9833792$$

$$S'_{3,4} = -0,3368697 \pm j0,4073290$$

Zeros: 4 at $S = \text{infinity}$

Denormalize with $\Omega_c = 1 \Rightarrow S_i = S'_i$

$$\text{LP} \rightarrow \text{HP}: S = \frac{\omega_I^2}{s}$$

Poles:

$$s_{1,2} = \frac{2}{T}(-0.2308 \pm j1.6267)$$

$$s_{3,4} = \frac{2}{T}(-1.96752 \pm j2.37905)$$

Zeros: 4 at $s = 0$

Transform to digital filter

$$z = \frac{1 + \frac{sT}{2}}{1 - \frac{sT}{2}}$$

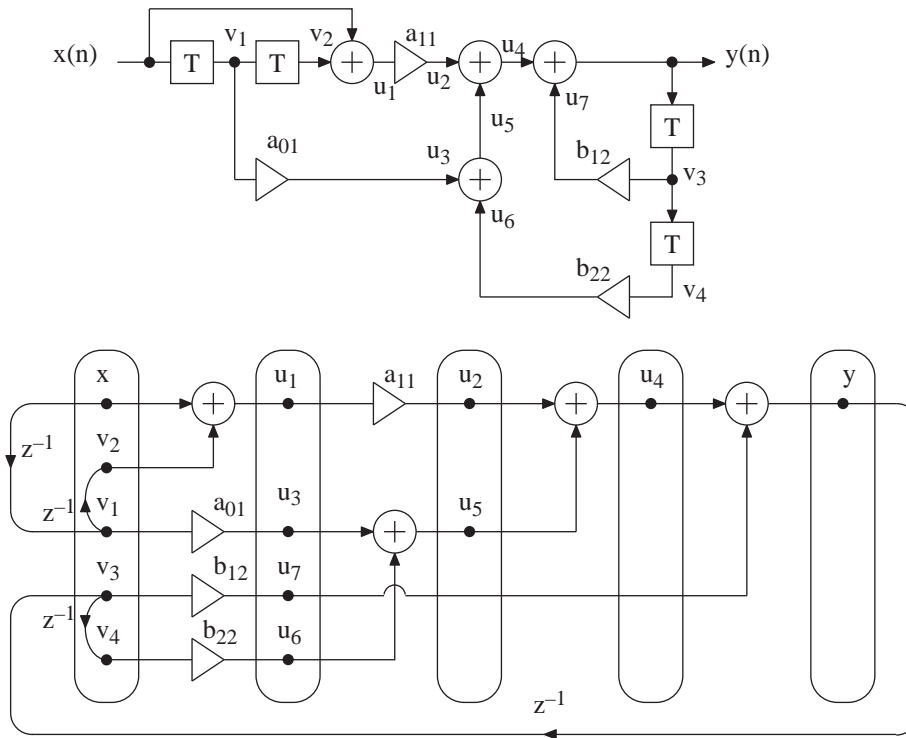
Poles:

$$z_{1,2} = -0.4084 \pm j0.781869$$

$$z_{3,4} = -0.589725 \pm j0.328915$$

Zeros: 4 at $z = 1$.

4 a)



b)

$$\begin{aligned}
 u_1 &= v_2 + x & u_7 &= u_4 + u_5 \\
 u_3 &= a_{01}v_1 & y &= u_4 + u_7 \\
 u_7 &= b_{12}v_3 & v_2 &= v_1 \\
 u_6 &= b_{22}v_4 & v_1 &= x \\
 u_2 &= a_{11}u_1 & v_4 &= v_3 \\
 u_5 &= u_3 + u_6 & v_3 &= y \\
 u_4 &= u_2 + u_5 & &
 \end{aligned}$$

c) $T_{\min} = T_{\text{mult}} + T_{\text{add}} = 1.1 \mu\text{s}$, $T_{\text{CP}} = T_{\text{mult}} + 3T_{\text{add}} = 1.3 \mu\text{s}$

5 a) $\sigma^2 = 2 \frac{Q^2}{12} \sum_{n=0}^{\infty} h_1^2(n)$

$$H_1(z) = \frac{1}{1 - b_1 z^{-1}} \Leftrightarrow h_1(n) = b_1^n u(n)$$

where $b_1 = 0.875$

$$\sum_{n=0}^{\infty} h_1^2(n) = \sum_{n=0}^{\infty} b_1^{2n} = \frac{1}{1 - b_1^2}$$

$$\sigma^2 = 2 \frac{Q^2}{12} \frac{1}{1-b_1^2} = 2 \frac{2^{-28}}{12} \frac{1}{1-0,875^2} \approx 2,65 \cdot 10^{-9}$$

b) $y(n)$ scaled by multiplying a_0 and a_2 by c .

$$c = \frac{1}{\sqrt{\sum_{n=0}^{\infty} h^2(n)}}$$

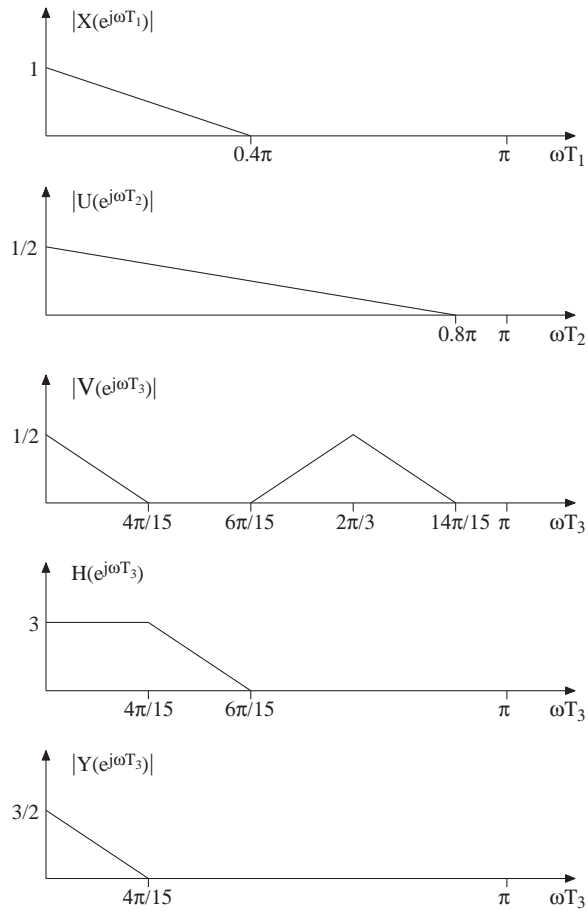
$$H(z) = \frac{a_0 + a_2 z^{-2}}{1 - b_1 z^{-1}} \Leftrightarrow h(n) = a_0 b_1^n u(n) + a_2 b_1^{n-2} u(n-2)$$

where $a_0 = 1, a_2 = -0.25, b_1 = 0.875$

$$h(n) = \begin{cases} 0, & n < 0 \\ a_0, & n = 0 \\ a_0 b_1, & n = 1 \\ \left(a_0 + \frac{a_2}{b_1^2}\right) b_1^n, & n > 1 \end{cases}$$

$$\begin{aligned} \sum_{n=0}^{\infty} h^2(n) &= a_0^2 + (a_0 b_1)^2 + \left(a_0 + \frac{a_2}{b_1^2}\right)^2 \sum_{n=2}^{\infty} b_1^{2n} \\ &= a_0^2 + (a_0 b_1)^2 + \left(a_0 + \frac{a_2}{b_1^2}\right)^2 \frac{b_1^4}{1-b_1^2} = 2,9 \end{aligned}$$

6)



7) $z = \frac{1 + \Psi}{1 - \Psi} \Rightarrow z_0 = 0.3609669, z_{1,2} = 0.3088943 + j 0.5644158$

$z_{3,4} = 0.2571957 + j 0.8712528$

$$H(z) = \left(\frac{-\alpha_0 z + 1}{z - \alpha_0} \right) \left(\frac{-\alpha_3 z^2 - \alpha_4 (1 - \alpha_3) z + 1}{z^2 - \alpha_4 (1 - \alpha_3) z - \alpha_3} \right) + \left(\frac{-\alpha_1 z^2 - \alpha_2 (1 - \alpha_1) z + 1}{z^2 - \alpha_2 (1 - \alpha_1) z - \alpha_1} \right)$$

$$= \left(\frac{-z_0 z + 1}{z - z_0} \right) \left(\frac{z_3 z_4 z^2 - (z_3 + z_4) z + 1}{z^2 - (z_3 + z_4) z + z_3 z_4} \right) + \left(\frac{z_1 z_2 z^2 - (z_1 + z_2) z + 1}{z^2 - (z_1 + z_2) z + z_1 z_2} \right)$$

$\alpha_0 = z_0 = 0,3609669$

$\alpha_1 = -z_1 z_2 = -0,4139809$

$\alpha_2 = \frac{z_1 + z_2}{1 - \alpha_1} = 0,4369144$

$\alpha_3 = -z_3 z_4 = -0,8252311$

$\alpha_4 = \frac{z_3 + z_4}{1 - \alpha_3} = 0,2818226$